

Implementation of the Huygens Absorbing Boundary Condition in Corner Regions

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Abstract—In recent years, three new absorbing boundary conditions (ABCs) have appeared in the literature, namely, the multiple absorbing surfaces, the reradiating boundary condition, and the Huygens ABC (HABC). The last is a generalization of the first two. The HABC mainly relies on the radiation of a field opposite to the outgoing field by means of surface currents. This paper focuses on the implementation of the HABC in the corner regions of computational domains. It is shown rigorously that the Huygens surface radiating the opposite field is not a normal Shelkunoff surface. Additional branches, called extensions, must be added in the corner regions.

Index Terms—Absorbing boundary condition (ABC), finite difference, finite-difference time-domain (FDTD) method, Huygens surface.

I. INTRODUCTION

TWO novel absorbing boundary conditions (ABCs) were presented independently some years ago in the literature: the multiple absorbing surfaces [1] and the reradiating boundary condition (rRBC) [2], [3]. Both rely on the same principle of canceling the outgoing field leaving a computational domain by means of equivalent currents that radiate a field equal in magnitude and opposite in sign to the field to be cancelled. This concept has been generalized and investigated in details in [4], where it is called the Huygens ABC (HABC).

As shown in [4], the HABC is equivalent to a traditional operator ABC. This is because the HABC concept cannot be implemented rigorously. More precisely, the required equivalent currents are not known on the boundary where they must be applied so that they are replaced with an estimate computed using an operator, for example, a Higdon operator [5], [6]. However, the HABC is not just another implementation of traditional operator ABCs [4]. First, it can be easily designed to absorb evanescent waves [4]. Second, it can be combined with such other ABCs as the PML ABC [4] or a real stretch of coordinates [7]. For these reasons, and since in addition it is simpler than the PML ABC, the HABC is a promising ABC that may challenge the well-established PML ABC in some problems, or can be used to improve the effectiveness of other ABCs.

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The papers [1]–[4] mainly present the principle and the theory of the proposed ABCs on a plane boundary assumed as of infinite extent. Although numerical experiments are reported in these papers in two or three dimensions, little attention is paid to the corners of the computational domain. By contrast, this paper focuses precisely on this question, which is a critical question since the equivalent currents that generate the opposite wave are not just a usual Huygens surface.

As stated by the Shelkunoff theorem, a Huygens surface is a closed and continuous surface that splits the physical space into two regions, in general, an interior region and an exterior region. Sources in one region can be replaced with equivalent currents that radiate the same field as the sources into the other region. Huygens surfaces are currently used with the finite-difference time-domain (FDTD) method either to generate an incident wave or for the near-to-far field transformation [8]. In Cartesian coordinates, they take the form of a parallelepiped. In the HABC [4] and in its special cases [1] and [2], [3], the situation is more complex. The reason is that the exact equivalent currents are replaced with estimates computed with an operator that is discontinuous at the corners of the HABC. If a normal Huygens surface is used, the radiated field is discontinuous so that spurious sources are produced in the corners. To overcome this problem, additional surfaces must be added to the Huygens surface. They extend from the corners or edges to the outer end of the computational domain. We call them the extensions of the Huygens surface.

In papers [1] and [2], [3], the need for extensions is not discussed and even mentioned. However, there is some evidence, as in Fig. 6 in [1] or in Table I in [3], that the Huygens planes were extended up to the outer boundary ending the domain. In other words, what we call here extensions were used in the numerical experiments. The rationale for the authors to do this is not reported. It may be a proper understanding of the corner problem or the simplification of the implementation suggested by a brief comment in [1]. Thus, in view of future developments of the HABC, there is a need for clarification and a proper theoretical justification of the implementation in the corners. This is the purpose of this paper. It shows in detail why the extensions are needed and reports some numerical experiments that demonstrate the impact of the extensions on the correctness of the computed results. The absence of extension has a limited impact on the results when the HABC is set closely to the outer boundary ending the domain, as in the experiments in [1]–[4], but when the space between the HABC and the outer boundary is large, as in the method [7], the presence of the extensions is primordial. Without extension, the computed results are strongly erroneous.

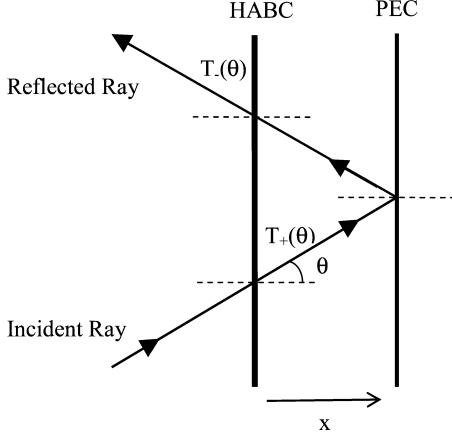


Fig. 1. A plane wave propagating toward an infinite wall HABC.

90 The content of this paper is as follows. Section II summarizes
 91 the properties of the HABC [4] on a wall boundary. Section III
 92 shows that the field radiated by a normal Huygens surface is
 93 discontinuous in the corner regions of the HABC. Section IV
 94 presents the extended HABC, which removes the discontinuity.
 95 Section V shows several experiments in the 2-D case to illustrate
 96 the theory. Finally, Section VI briefly illustrates the effectiveness
 97 of the combination of the HABC with a stretch of coordinates
 98 [7] to solve some typical open problems of electromagnetic
 99 compatibility.

100 II. HABC ON A WALL BOUNDARY

101 Let us consider a plane HABC placed in front of a plane PEC,
 102 with an incident wave propagating at an angle of incidence θ .
 103 This is a 3-D problem depicted in Fig. 1 looking along the plane
 104 of incidence. From [4], the estimate of the field at the HABC
 105 location can be written as follows:

$$\begin{aligned} \tilde{U}(x_{\text{HABC}}, t) &= \sum_{k=1}^M a_k U(x_{\text{HABC}} - \delta x_k, t - \delta t_k) \\ &+ \sum_{k=M+1}^N a_k \tilde{U}(x_{\text{HABC}}, t - \delta t_k) \end{aligned} \quad (1)$$

106 where U is either the E or H field, and \tilde{U} is the estimate of U .
 107 The first term is a linear function of U at M interior locations
 108 $x_{\text{HABC}} - \delta x_k$ and previous times $t - \delta t_k$. The second term is
 109 a combination of $N - M$ estimates at previous times $t - \delta t_k$.
 110 When an incident wave $U_{i+}(t)$ comes from the left-hand side,
 111 from [4], the wave $U_{t+}(t)$ transmitted by the HABC, i.e., the
 112 field to the right of the HABC if the PEC were absent, is the
 113 derivative on time of the incident wave:

$$U_{t+}(t) = T_+(\theta) \frac{\partial U_{i+}(t)}{\partial t} \quad (2)$$

114 with

$$T_+(\theta) = \frac{1}{1 - \sum_{k=M+1}^N a_k} \sum_{k=1}^N a_k \left(\frac{\delta t_k - \cos \theta \delta x_k}{c} \right) \quad (3)$$

where c is the speed of light. We assume that the numerical
 115 technique is the FDTD method [8] so that the shifts in space and
 116 time δx_k and δt_k are multiples of the steps on space and time
 117 Δx and Δt , respectively. We use the first-order Higdon operator
 118 with which the estimate reads:
 119

$$\begin{aligned} \tilde{U}^{n+1}(I_{\text{HABC}}) &= U^n(I_{\text{HABC}} - 1) + w U^{n+1}(I_{\text{HABC}} - 1) \\ &- w \tilde{U}^n(I_{\text{HABC}}) \end{aligned} \quad (4)$$

where n is the index on time, I is the index on space in x
 120 direction, and
 121

$$w = \frac{c\Delta t - \Delta x}{c\Delta t + \Delta x}. \quad (5)$$

The estimation (4) is a special case of (1) with $M = 2$ and $N =$
 122 3. Using (1), (3), and (4), the coefficient $T_+(\theta)$ of the Higdon
 123 operator is obtained as follows:
 124

$$\begin{aligned} T_{+\text{Hig}}(\theta) &= \frac{1}{1+w} \left[\frac{(1-w)\Delta t - \cos \theta (1+w)\Delta x}{c} \right] \\ &= \frac{(1 - \cos \theta) \Delta x}{c}. \end{aligned} \quad (6)$$

Consider now a wave $U_{i-}(t)$ propagating from the right-hand
 125 side of the HABC, such as the reflected ray in Fig. 1. From [4],
 126 the wave $U_{t-}(t)$ transmitted to the left-hand side of the HABC
 127 is the integral on time of $U_{i-}(t)$:
 128

$$U_{t-}(t) = T_-(\theta) \int_{-\infty}^t U_{i-}(t') dt' \quad (7)$$

with

$$T_-(\theta) = \frac{1 - \sum_{k=M+1}^N a_k}{\sum_{k=1}^N a_k (\delta t_k + \cos \theta \delta x_k / c)} \quad (8)$$

or in the special case of the Higdon operator:

$$T_{-\text{Hig}}(\theta) = \frac{1+w}{[(1-w)\Delta t + \cos \theta (1+w)\Delta x / c]} = \frac{c}{(1 + \cos \theta) \Delta x}. \quad (9)$$

For the problem in Fig. 1, the incident wave is differentiat-
 131 ed by the HABC and transmitted with the coefficient $T_+(\theta)$,
 132 reflected from the PEC with the coefficient -1 , and then inte-
 133 grated by the HABC and transmitted back into the interior
 134 domain with the coefficient $T_-(\theta)$. The net result is the apparent
 135 reflection:
 136

$$R(\theta) = -T_-(\theta)T_+(\theta) \quad (10)$$

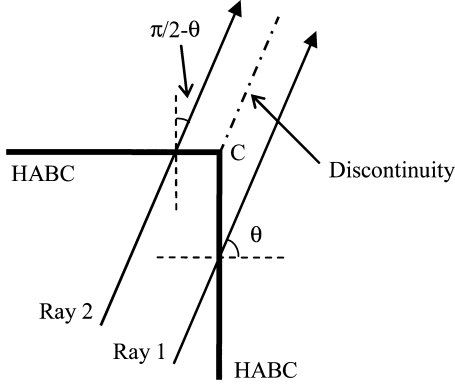


Fig. 2. Discontinuity of the transmitted wave at the corner of a 2D HABC.

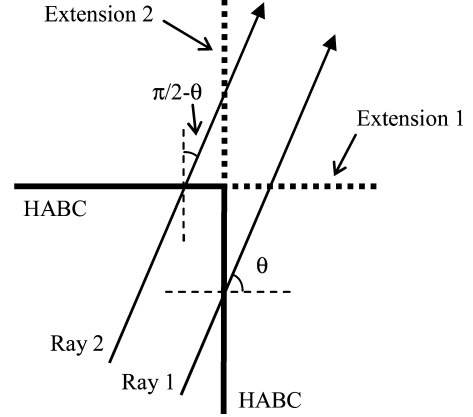


Fig. 3. The extensions of the HABC at the corners in the 2D case.

137 where a term corresponding to the free space propagation be-
 138 tween HABC and PEC has been omitted. For pure traveling
 139 waves, this term is a phase term of modulus one. For evanescent
 140 waves, it would equal the natural decay of the wave. As shown
 141 in [4], the reflection (10) equals, rigorously, the reflection ob-
 142 tained by enforcing the estimate (1) as a boundary condition.
 143 This is easily verified in the case of the Higdon operator, where
 144 using (6), (9), and (10), we obtain $R(\theta) = -(1 - \cos\theta)/(1 +$
 145 $\cos\theta)$, which is the well-known reflection from the first-order
 146 Higdon operator [5], [6]. Therefore, the ABC composed of a
 147 HABC surface and a PEC surface is equivalent, rigorously, to
 148 the corresponding operator ABC.

149 III. HABC AT THE EDGES AND CORNERS OF A 150 COMPUTATIONAL DOMAIN

151 Let us first consider the 2-D case depicted in Fig. 2, where
 152 an HABC without PEC behind it radiates the wave opposite to
 153 the incident wave. Assume that the outgoing wave is a plane
 154 wave at incidence θ with respect to the vertical in Fig. 2, and
 155 consider two rays that strike the HABC to the right and to the
 156 left of the corner C. Then the incidence on the Huygens surface
 157 is θ for ray 1 and $\pi/2 - \theta$ for ray 2. The transmitted waves
 158 corresponding to the two rays are the derivative on time of
 159 the incident wave, multiplied with the coefficients $T_+(\theta)$ and
 160 $T_+(\pi/2 - \theta)$, respectively, where T_+ is given by (3) in general
 161 or (6) in the special case of the Higdon operator. It is obvious
 162 from (3) or (6) that the magnitudes of the two transmitted rays
 163 are different, except in the case where $\theta = \pi/4$. This is true for
 164 all rays as long as they strike the HABC on the two sides of the
 165 corner C, and whatever may be their distance from C. Thus, the
 166 transmitted field is discontinuous behind the HABC. The wave
 167 is no longer a plane wave. Its magnitude varies in the direction
 168 perpendicular to the propagation. This also occurs in 3-D at the
 169 edges of the HABC, which are similar to 2-D corners, and at
 170 the 3-D corners, where there are three different transmission
 171 coefficients. In both 2- and 3-D, such discontinuities produce
 172 additional components to the field, i.e., they act as spurious
 173 sources.

174 The discontinuity in the transmitted wave originates in the
 175 estimate of the opposite field radiated by the Huygens surface.
 176 If the exact outgoing field was used as equivalent current, the

177 radiated field would be exact. With the HABC, the unknown
 178 outgoing field is replaced with an estimate computed using an
 179 operator that depends on the incidence on the Huygens surface.
 180 At the corner, the operator and then the estimates (3) or (6) are
 181 discontinuous simply because the incidence angle experiences
 182 an abrupt change, from θ to $\pi/2 - \theta$ in the 2-D case in Fig. 2.
 183 This results in a discontinuity in the radiated field that produces
 184 a nonphysical source that may be very large, as demonstrated
 185 by a numerical experiment in the Section V.

186 IV. EXTENDED HABC

187 A simple modification of the Huygens surface permits the
 188 discontinuity in the transmitted field to be removed. Let us first
 189 consider the 2-D case. The modification is depicted in Fig. 3.
 190 It consists of extending the HABC surfaces (reduced to lines in
 191 2-D) up to infinity in theory or up to the end of the computational
 192 domain in practice. The equivalent currents on the extensions are
 193 enforced as on the normal HABC surfaces. For instance, the
 194 equivalent currents on extension 1 are computed using the same
 195 operator as on the normal horizontal HABC and are enforced
 196 by means of the same modifications to the Maxwell equations.

197 With the extended HABC in Fig. 3, ray 1 crosses the verti-
 198 cal HABC, where it is differentiated on time and transmitted
 199 with coefficient $T_+(\theta)$. Then it crosses extension 1, where it is
 200 differentiated another time and transmitted with $T_+(\pi/2 - \theta)$.
 201 Similarly, ray 2 crosses the horizontal HABC, where it is differ-
 202 entiated and transmitted with $T_+(\pi/2 - \theta)$, and then crosses
 203 extension 2, where it is differentiated and transmitted with
 204 $T_+(\theta)$. In the region in-between the two extensions, the two
 205 transmitted waves are the second derivative of the incident wave
 206 and their magnitudes are equal to $T_+(\theta) T_+(\pi/2 - \theta)$. Thus, in
 207 this region, the discontinuity of the field is removed and the field
 208 remains a plane wave.

209 In summary, the space outside the HABC is split into three
 210 regions. Two regions, where the field is the first derivative on
 211 time, separated by the region, where the field is the second
 212 derivative on time. The field is a true plane wave in each region.
 213 As verified by numerical experiments in Section V, no spurious
 214 source is present.

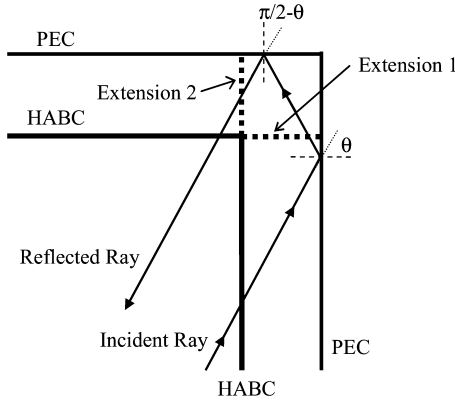


Fig. 4. Reflection of a plane wave from the corner of an extended HABC.

215 Fig. 3 and the aforementioned derivations address the 2-D
 216 case. In 3-D, there are edge regions similar to the 2-D
 217 corners. At the 3-D corners, by extending the three Huygens
 218 planes to infinity, the space outside the HABC is split into
 219 seven regions. It can be seen that the field in all the
 220 regions is a true plane wave, without discontinuity of the
 221 transmission coefficient. In three regions, the field is the
 222 first derivative of the incident wave, in three others it is
 223 the second derivative, and in the last it is the third
 224 derivative. In this seventh region, the transmission is the
 225 product of three coefficients (3) or (6) and is the same for
 all the rays so that the continuity of the field is ensured.

226 Consider now the reflection of a plane wave from the corner
 227 region when a PEC is present behind the HABC, as depicted
 228 in Fig. 4. The ray shown in Fig. 4 is differentiated and
 229 transmitted by the vertical HABC with the coefficient $T_+(\theta)$,
 230 reflected from the PEC, differentiated and transmitted by the
 231 horizontal extension 1 with $T_+(\pi/2 - \theta)$, reflected from the
 232 horizontal PEC, integrated and transmitted by extension 2
 233 with $T_-(\theta)$, and finally integrated and transmitted by the
 234 horizontal HABC with $T_-(\pi/2 - \theta)$. The net reflection reads:

$$R_{\text{EXTcorner}} = T_+(\theta)T_+ \left(\frac{\pi}{2 - \theta} \right) T_-(\theta)T_- \left(\frac{\pi}{2 - \theta} \right). \quad (11)$$

235 Using (10), this can be rewritten as follows:

$$R_{\text{EXTcorner}}(\theta) = R_{\text{ABC}}(\theta)R_{\text{ABC}} \left(\frac{\pi}{2 - \theta} \right) \quad (12)$$

236 where $R_{\text{ABC}}(\theta)$ and $R_{\text{ABC}}(\pi/2 - \theta)$ are the reflections
 237 from the operator ABC on the vertical and horizontal boundaries,
 238 respectively. From this, the reflection from the corner of the
 239 HABC is identical to the reflection from the corner of the
 240 domain bounded by the operator ABC relying on the same
 241 operator as the HABC. Obviously this is not true in the
 242 absence of extensions, since then the coefficients $T_+(\pi/2 - \theta)$
 243 and $T_-(\theta)$ are missing in (11) so that the reflection
 244 differs from the reflection from the corner of an operator
 245 ABC (12). The presence of the extensions permits the
 246 discontinuity to be removed in the corner, and renders the
 247 HABC equivalent, rigorously, to the corresponding operator
 ABC.

248 Fig. 4 addresses the 2-D case. In 3-D, the three Huygens
 249 planes are extended up to the PEC. The reflection from the

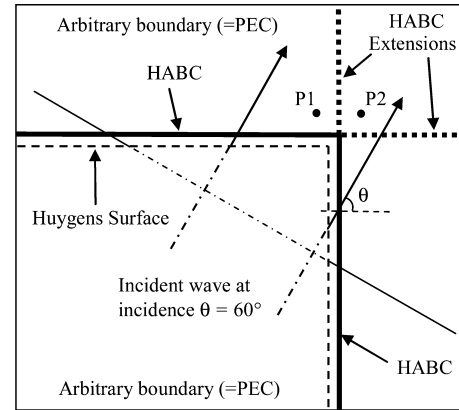


Fig. 5. Numerical experiments with a plane wave striking a corner of the HABC. The spatial and time steps of the 2D FDTD domain are 5 cm and 100 ps, respectively.

HABC corners is then the product of six coefficients, which
 250 is the generalization of (11) to 3-D. As in 2-D, this reflection
 251 equals that of the corresponding operator ABC. 252

253 The implementation of the HABC extensions in a computer
 254 code is simple. Since the equivalent sources on the extensions
 255 are identical to those on the normal HABC surfaces, the only
 256 thing to do is a change of the limits of the loops, where the
 257 equivalent currents are enforced. This is apparently what was
 258 done with the rRBC from the Table I in [3].

259 In principle, the extended HABC could be used with other
 260 outer boundaries than the PEC, so as to combine the HABC
 261 with another ABC [4]. At least with operator ABCs, which use
 262 only nodes located in the direction perpendicular to the ABC.
 263 This is the case with Higdon operators, but not with Engquist-
 264 Majda ABC [9], which uses FDTD nodes in the transverse
 265 direction. With the PML ABC, we think that the extension of
 266 the HABC is possible as well by using extensions up to the
 267 PEC ending the PML.

268 Only homogeneous waves have been considered previously.
 269 However, the results and conclusions in [4] also apply to non-
 270 homogeneous waves. Thus, the derivations and conclusions
 271 about the corner regions are also valid for evanescent waves,
 272 i.e., the extensions are also needed with evanescent waves.
 273 This is illustrated in Section V by an experiment with a
 274 scattering structure surrounded with strongly evanescent fields.

275 V. NUMERICAL EXPERIMENTS 275

276 This section reports two experiments, which validate the theory
 277 in the previous section, and one experiment, which demonstrates
 278 that the extension of the HABC is of primordial importance
 279 for further developments of the HABC technique. The
 280 experiments were performed in the 2-D case, which permits an
 281 easy generation of incident plane waves.

282 A. Continuity of the Field Behind the Extended HABC 282

283 In this experiment depicted in Fig. 5, a plane wave strikes
 284 the corner of a 2-D HABC. The wave is generated by means of a
 285 Huygens surface [8] placed close to the HABC. Both surfaces

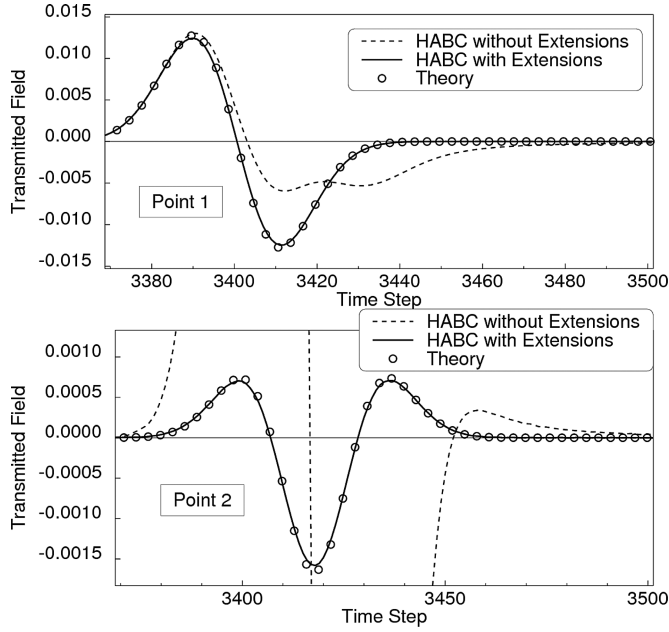


Fig. 6. The field transmitted behind the HABC in the absence of PEC, at P1 (upper part) and P2 (lower part) in Fig. 5. Comparison of the theoretical field with the FDTD fields computed with and without HABC extensions.

286 are truncated by nonphysical conditions so that the calculation
 287 is only valid for a clear time which is about 350 time steps at
 288 the corner region with the larger than 2000 cells domain used in
 289 the experiment. The incidence is 60° with respect to the vertical
 290 ABC. The incident wave is a Gaussian pulse 15 time steps in
 291 width. The FDTD domain is large enough behind the HABC
 292 in order to ensure that the reflection from its outer boundary is
 293 not viewed in the corner region for the whole duration of the
 294 calculation.

295 The calculations have been performed using the HABC with
 296 and without extensions. The results are shown in Fig. 6 at the
 297 two locations denoted as P1 and P2 in Fig. 5.

298 At point P1, the wave should be the derivative of the incident
 299 wave multiplied with the coefficient $T_{+Hig}(\pi/2 - \theta)$. This
 300 theoretical prediction is plotted in the upper part of Fig. 6. The
 301 extended HABC result agrees very well with the theory. Without
 302 extension, the field differs from the prediction because of the
 303 discontinuity of the transmission by the HABC.

304 At point 2, the field with the extended HABC should be the
 305 second derivative of the incident field multiplied with $T_{+}(\pi/2 - \theta)$
 306 and $T_{-}(\theta)$. The lower part of Fig. 6 compares this theoretical
 307 prediction with the FDTD results. The agreement is excellent
 308 with the extended HABC. Conversely, without extension, the
 309 field is quite different because of the discontinuity and because
 310 it remains proportional to the first derivative so that its magnitude
 311 is about that at P1.

312 *B. Comparison of the HABC With an Operator ABC*
 313 *in a Corner Region*

314 In this experiment depicted in Fig. 7, a PEC is present 20
 315 FDTD cells behind the HABC. As in Fig. 5, the incident wave
 316 is generated by a Huygens surface close to the HABC so that

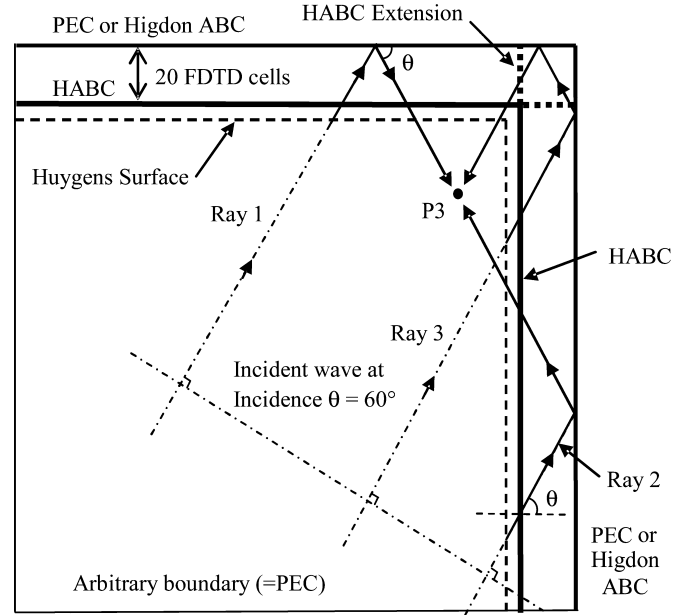


Fig. 7. Numerical experiments with a plane wave striking a corner of the HABC, with a PEC behind the HABC.

TABLE I
 THE THREE CALCULATIONS WITH THE DOMAIN IN FIG. 7

Calculation	Outer boundary	HABC	HABC extension
1	PEC	yes	no
2	PEC	yes	yes
3	Higdon ABC	no	no

only the reflected field is present in the interior domain at the
 observation point P3. Three calculations were performed whose
 calculation settings are summarized in Table I. The first one
 with the PEC and the HABC without extension, the second one
 with the PEC and the HABC with extensions, and the third one
 without HABC and with the PEC replaced with the Higdon
 operator ABC. The three reflected fields at point P3 are plotted
 in Fig. 8, with in addition to the theoretical reflection from a
 Higdon operator ABC.

We can observe three pulses in Fig. 8. They correspond to the
 three rays represented in Fig. 7. Ray 2 is the reflection from the
 vertical boundary with incidence θ , ray 1 from the horizontal
 boundary with incidence $\pi/2 - \theta$, and ray 3 from the corner. For
 the Higdon operator ABC and $\theta = 60^\circ$, the corresponding reflections
 $R(\theta)$, $R(\pi/2 - \theta)$, and $R(\theta) R(\pi/2 - \theta)$ are 0.3333, 0.0718,
 and 0.0239, respectively. These reflections have been used along
 with the differences between the ray paths to compute the theo-
 retical reflection plotted in Fig. 8. As observed, the Higdon
 operator result and the extended HABC result agree very well
 with the expected theoretical reflection. The extended HABC is
 equivalent, rigorously, to the Higdon ABC. Conversely, with-
 out extensions, the HABC yields a strongly different result, the
 reflection from the corner (ray 3, third pulse) is one order of

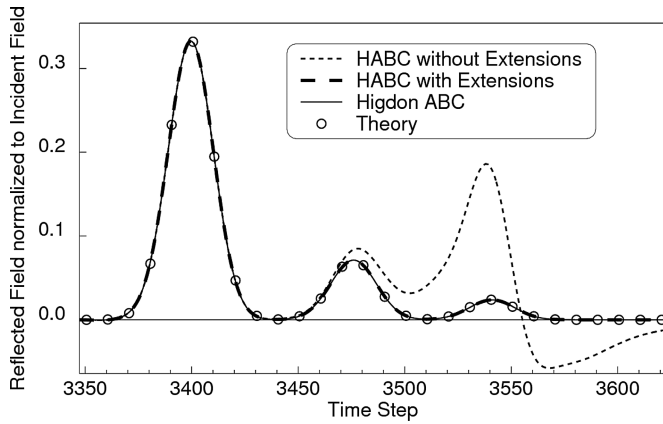


Fig. 8. The reflected field at point P3 in Fig. 7. Comparison of the theoretical field with the results of the three FDTD calculations defined in the Table I.

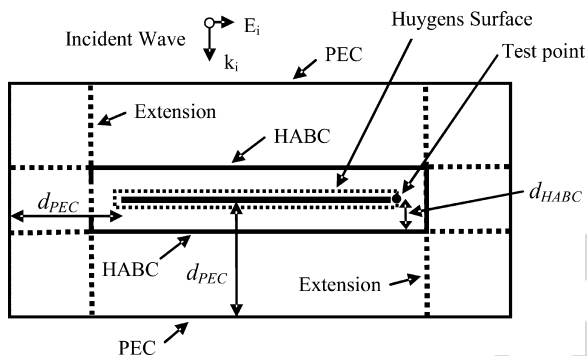


Fig. 9. The 2D FDTD domain for the experiments with a plane wave striking a 2D 300-cell long thin plate. The FDTD steps equal 5 cm and 100 ps. The incident electric field is parallel to the PEC plate and is generated by a normal Huygens surface set 2 FDTD cells from the plate. The HABC is d_{HABC} from the plate and the outer PEC is d_{PEC} from the plate.

340 magnitude larger than the correct reflection, because of the spu-
 341 rious source produced by the discontinuity of the transmitted
 342 wave.

343 C. Experiment With a Scattering Structure

344 A 300-cell-long 2-D PEC plate of zero thickness is struck by
 345 an incident plane wave propagating downward (see Fig. 9). The
 346 incident waveform is a unit step with a rise time of 10 time steps.
 347 An HABC surface and a PEC are placed at d_{HABC} and d_{PEC}
 348 from the plate, respectively. For the experiments reported in
 349 Fig. 10, the HABC surface is placed at various distances d_{HABC}
 350 from the PEC structure, ranging from 10 to 898 FDTD cells, and
 351 the PEC is 900 cells from the structure in all the calculations.
 352 Fig. 10 shows the electric field normal to the surface at the end
 353 of the plate. The HABC without extensions is used in the upper
 354 part of Fig. 10, and the extended HABC in the lower part.

355 As can be observed in Fig. 10, the effect of the extension
 356 is very important when the separation between the scattering
 357 structure and the HABC becomes small. Even with $d_{HABC} =$
 358 10 cells, which is only 1/30 of the structure size, the extended
 359 HABC yields results superimposed on the solution computed
 360 with the HABC placed 900 cells away, i.e., almost superimposed
 361 on the exact solution. This means that the Higdon operator

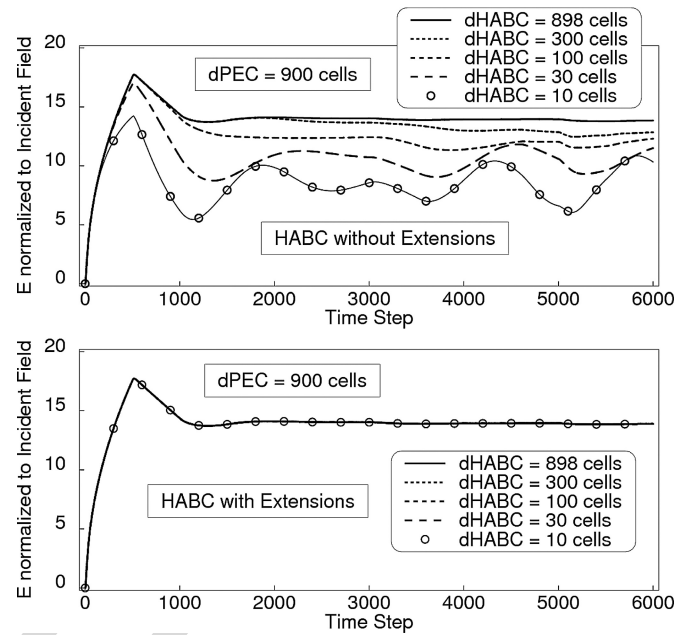


Fig. 10. Comparison of the HABC without extension (upper part) and with the extensions (lower part) when a plane wave strikes a PEC object. The outer PEC is 900 cells from the object and the HABC is placed various distances from the object.

362 implemented as an HABC can very well absorb the traveling
 363 waves even when it is quite close to the scattering structure.
 364 Obviously, the evanescent waves are not absorbed by the HABC
 365 based on the Higdon operators [4]. In the experiments in Fig. 9,
 366 they decrease in the large space surrounding the HABC so that
 367 their apparent reflection is negligible.
 368

368 VI. EFFECTIVE ABC FOR ELECTROMAGNETIC 369 COMPATIBILITY PROBLEMS

370 Typical problems of electromagnetic compatibility consist of
 371 computing the field on the surface or in the vicinity of a PEC
 372 object struck by an incident wave. In such problems, the scat-
 373 tered field is composed of traveling waves at high frequency
 374 and evanescent waves at low frequency, with a transition about
 375 the fundamental resonance of the object. As shown above, an
 376 HABC with the Higdon operator effectively absorbs the travel-
 377 ing waves, even if it is close to the object. However, a large
 378 domain is still needed outside the HABC for the natural decrease
 379 of the evanescent waves.

380 To reduce the exterior domain, several ways can be imag-
 381 ined. One is the introduction of an operator designed for the
 382 evanescent waves, used either as a traditional operator ABC or
 383 in the form of an additional HABC. This was tested success-
 384 fully in waveguides [4]. Another idea was introduced in [7].
 385 It consists in keeping a large physical domain, as in Fig. 10,
 386 but with a strongly stretched FDTD mesh, so as to reduce the
 387 overall number of FDTD cells. This is possible because only
 388 the low frequency evanescent waves must be absorbed outside
 389 the HABC. Since their characteristic length of decrease is of the
 390 order of the structure size, use of quite large FDTD cells can
 391 be envisaged. This has been confirmed by experiments such as that

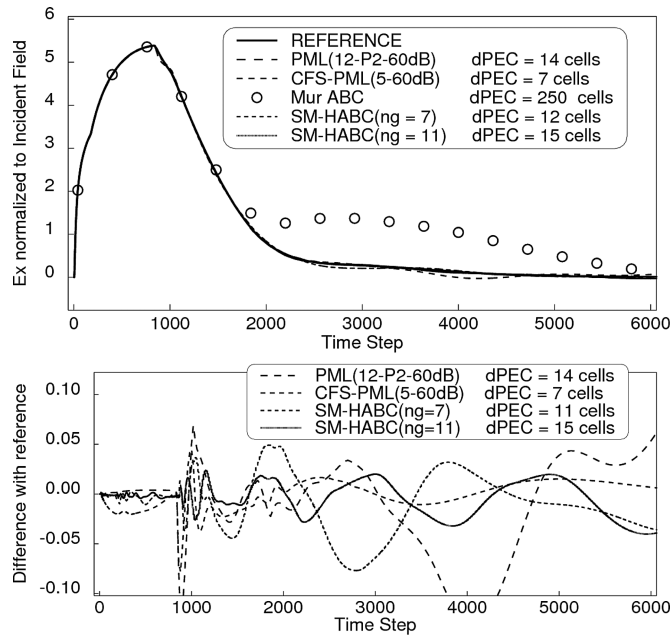


Fig. 11. Electric field at the corner on the surface of a 2D 500×50 PEC parallelepiped computed with various absorbing boundary conditions. The lower part is the difference of the computed result with the reference solution.

in [7]. Thus the combination of an HABC for the absorption of the traveling waves with a stretched mesh for the absorption of the evanescent waves forms an effective ABC. We call it the Stretched Mesh HABC (SM-HABC).

A comparison of the SM-HABC with other ABCs is provided in Fig. 11 with a 2D canonical case. The incident wave is a double exponential of the form $\exp(-t/t_f) - \exp(-t/t_r)$ where $t_r = 1$ ns and $t_f = 100$ ns. The 2D object is a 25 m long parallelepiped of 500×50 FDTD cells of size 5 cm.

The settings of the SM-HABC calculations are similar to those in Fig. 9, with the 500×50 -cell object surrounded with a HABC placed 3 cells from it, and with a large exterior domain filled with strongly stretched cells. The separation d_{PEC} between the object and the outer PEC is 75 m, that is 3 object sizes. The stretch of the mesh is a geometrical expansion that begins 4 cells from the object. Two cases are reported in Fig. 11. In the first one, the mesh is stretched upon $n_g = 7$ cells up to the outer PEC, with ratio $g = 2.68$. This means that the separation d_{PEC} , which corresponds to a physical distance of 1500 cells of 5 cm, is filled with only 11 FDTD cells. In the second case, the mesh is stretched upon $n_g = 11$ cells, with $g = 1.82$, so that the actual separation is $d_{PEC} = 15$ cells.

Two calculations were performed with a PML ABC placed 2 FDTD cells from the object. The first one used a 12 cell thick PML with a polynomial conductivity of power 2 and the normal stretching factor. This PML is not optimum in the sense of [10], but it is probably representative of the PMLs employed by most users. The second one used the best PML which is the CFS-PML optimized for 3D wave-structure interactions [11], [12]. It is only 5 cells in thickness. Finally another result in Fig. 11 was computed with the second order Engquist-Majda ABC [9] placed 250 cells from the object.

It is clearly seen in Fig. 11 that the SM-HABC can challenge the PML ABC. Despite the extremely large cells used to fill the large domain, the accordance with the reference solution is similar to the accordance of the PML ABC. More precisely, the lower part of the figure shows that the two SM-HABCs outperform the normal 12-cell PML. Also, the magnitude of the error with the SM-HABC and $ng = 11$ cells is very close to that observed with the optimum CFS-PML. The overall number of FDTD cells is slightly larger with the SM-HABC than with the optimum CFS-PML, but the cost of one cell of vacuum is smaller than the cost of one cell of PML. Thus the computational costs of the two ABCs are roughly similar. However, the SM-HABC has a significant advantage in comparison with the CFS-PML. Its implementation is far simpler. This is an attractive feature.

The same conclusions also hold in the 3D case. This will be demonstrated in a forthcoming paper devoted to experiments with realistic 3D scattering objects.

VII. CONCLUSION

The implementation of the HABC in the corner regions of 2D or 3D computational domains has been analysed in details. We have shown that the Huygens surface must be extended up to the surrounding PEC. The extensions that may seem in discordance with the equivalence theorem are necessary because of the replacement of the exact outgoing field with an estimate that is discontinuous at the corners of the HABC. The extensions remove the spurious sources produced by the discontinuity, and render the HABC rigorously equivalent to an operator ABC.

With the extensions the HABC is a highly effective ABC for the absorption of the travelling waves. By combining the HABC with a strategy to absorb the evanescent waves, as done in this paper with the SM-HABC, highly effective ABCs can be constructed for the solution of problems of electromagnetism, especially in the field of electromagnetic compatibility.

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